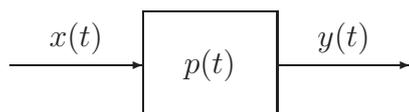


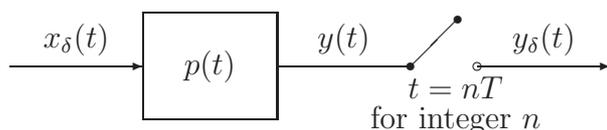
### Sample Problems for the Tenth Quiz

1. (Alternative proof for Nyquist Criterion)

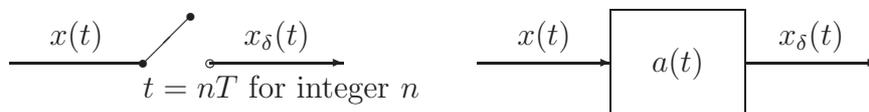
- (a) Find the LTI filter design with impulse response  $p(t)$  that guarantees  $y(t) = x(t)$ .



- (b) Find the LTI filter design with impulse response  $p(t)$  that guarantees  $y_\delta(t) = x_\delta(t)$ , where  $x_\delta(t) = \sum_{k=-\infty}^{\infty} x(nT)\delta(t - nT)$ .



2. Is a sampler linear? Can we find an equivalent linear time-invariant filter with impulse response  $a(t)$  such that  $x_\delta(t) = x(t) \star a(t)$ ?



Hint:  $x_\delta(t) = \sum_{n=-\infty}^{\infty} x(nT)\delta(t - nT)$ .

3. (a) Show that the time-averaged PSD of

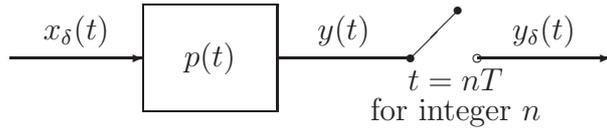
$$s(t) = \sum_{n=-\infty}^{\infty} a_n g(t - nT)$$

for WSS  $\{a_n\}_{n=-\infty}^{\infty}$  is

$$\overline{\text{PSD}}_s(f) = |G(f)|^2 \left( \frac{1}{T} \sum_{n=-\infty}^{\infty} R_a(n) e^{-i2\pi f n T} \right)$$

where  $G(f)$  is the Fourier transform of  $g(t)$  and  $R_a(n) = E[a_{n+m}a_n^*]$  is the autocorrelation function of WSS  $\{a_n\}_{n=-\infty}^{\infty}$ .

- (b) Use (a) to find the time-averaged PSD of  $x_\delta(t)$  and  $y(t)$  below, where  $x_\delta(t) = \sum_{k=-\infty}^{\infty} a_k \delta(t - kT)$  and  $\{a_n\}_{n=-\infty}^{\infty}$  WSS. Note that here  $p(t)$  is a general LTI filter and may not satisfy the Nyquist Criterion (and is possibly a design of correlative level coding).



- (c) Continue from (b). We note that  $y_\delta(t) = \sum_{n=-\infty}^{\infty} c_n \delta(t - nT)$ , where  $c_n = y(nT)$ . If  $c_n = a_n + a_{n-1}$  as the duo-binary correlative level coding, is  $\{c_n\}_{n=-\infty}^{\infty}$  also WSS? Represent the time-averaged PSD of  $y_\delta(t)$  in terms of that of  $x_\delta(t)$ .
4. From the slides, we know that the error propagation of correlative level coding can be eliminated by pre-coding. Two examples are given in our lectures, which are:

$$\text{Duo-binary} \begin{cases} b_k = \tilde{b}_k \oplus \tilde{b}_{k-1} \\ a_k = 2\tilde{b}_k - 1 \\ c_k = a_k + a_{k-1} \end{cases} \\ \Rightarrow b_k \text{ can be fully determined by } c_k.$$

$$\text{Modified Duo-binary} \begin{cases} b_k = \tilde{b}_k \oplus \tilde{b}_{k-2} \\ a_k = 2\tilde{b}_k - 1 \\ c_k = a_k - a_{k-2} \end{cases} \\ \Rightarrow b_k \text{ can be fully determined by } c_k.$$

Give one pre-coding scheme for

$$\begin{cases} b_k = ?? \\ a_k = 2\tilde{b}_k - 1 \\ c_k = a_k + 2a_{k-1} + a_{k-2} \end{cases} \Rightarrow b_k \text{ can be fully determined by } c_k.$$

Hint:  $b_k = (\alpha \tilde{b}_k) \oplus (\beta \tilde{b}_{k-1}) \oplus (\gamma \tilde{b}_{k-2})$  for some  $(\alpha, \beta, \gamma) \in \{0, 1\}^3$ . There are eight possible candidates for precoding design.