

Code Construction for Gauss-Markov Fading Channel

Prepared by Chin-Hui Cheng

Advisory by Prof. Po-Ning Chen

Institute of Communications Engineering

National Chiao-Tung University

Hsinchu, Taiwan 300, R.O.C.

June 18, 2005

Outline

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Introduction

- Simulated annealing has been applied to code design.
- In 2002, Skoglund, Giese and Parkvall designed a block code and the performance over the block Rayleigh-fading channel was excellent [7].
- Since the channel coefficients are changing during the period of transmission of codeword in practice, we try to design code to adjust well to Gauss-Markov channels.

Simulated Annealing for Code Design

- "Anneal" means to heat and then to cool steel or glass. The cooling stage is slowly so that the potential energy stored in the molecular configuration can be minimized
- A code structure is similar to the molecular constitution in the aspect of the analogy between good distance properties of a code and low potential energy of a molecular configuration.

- Choose initial temperature $T = T_0$, code C
- ```
Do{
 Do{
 Choose C', a random perturbation of C
 Let $\Delta e = \text{energy}(C') - \text{energy}(C)$
 If ($\Delta e < 0$) then $C \leftarrow C'$
 Else with probability $\exp(-\Delta e/T)$ $C \leftarrow C'$
 }
 Until (several energy drops or too many iterations)
 Lower temperature $T \leftarrow \alpha T$
}
```
- Until (stable code configuration is obtained or running time is up )

## Channel Characteristic

- For a time-varying channel with noise and inter-symbol interference, the received data at time  $k$  is

$$r_k = \mathbf{a}_k^T \mathbf{h}_k + n_k,$$

where  $\mathbf{a}_k^T = [a_k, a_{k-1}, \dots, a_{k-M+1}]$  and  $M$  is the channel length indicating how far the past transmitted bits affect the received bit in this time instance.  $\mathbf{h}_k$  is a complex column vector with  $M$  channel impulse response coefficients at time  $k$ , and  $n_k$  is a Gaussian complex noise with mean zero, variance  $\sigma^2$ .

- For Gauss-Markov channel,  $\mathbf{h}_k = \alpha \mathbf{h}_{k-1} + \mathbf{v}_k$  at time  $k$ , where  $\mathbf{v}_k$  is a complex white Gaussian with mean  $\mathbf{d}$  and covariance  $\mathbf{C}$ .

- Let  $\mathbf{H} = [\mathbf{h}_1, \mathbf{h}_2, \dots, \mathbf{h}_N]$  be the matrix of channel coefficient vectors representing by columns, and let  $\mathbf{A} = [\mathbf{a}_1, \dots, \mathbf{a}_N]^T$  represent the matrix of transmitted data representing by rows. Also let  $\mathbf{r} = [r_1, \dots, r_N]^T$  represent the received vector, while  $\mathbf{n} = [n_1, \dots, n_N]^T$  be the vector of noise from time 1 to time  $N$ .
- Assuming the mean of  $\mathbf{v}_k$  is zero and initial channel coefficient  $\mathbf{h}_0$  is known, we can derive the likelihood function of Gauss-Markov channel as

$$\begin{aligned}
 f(\mathbf{r}|\mathbf{A}) &= E[f(\mathbf{r}|\mathbf{H}, \mathbf{A})] \\
 &= \int_{\mathbf{H}} f(\mathbf{r}|\mathbf{H}, \mathbf{A}) f(\mathbf{H}) d\mathbf{H} \\
 &= \int_{\mathbf{H}} \left( \prod_{k=1}^N e^{-\frac{|r_k - \mathbf{a}_k^T \mathbf{h}_k|^2}{\sigma^2}} \right) f(\mathbf{h}_1) \prod_{k=2}^N f(\mathbf{h}_k | \mathbf{h}_{k-1}) d\mathbf{H}.
 \end{aligned}$$

- After marginalizing the channel coefficient distribution,

$$f(\mathbf{r}|\mathbf{A}) = \prod_{k=1}^n e^{\mathbf{q}_k^H \mathbf{G}_k \mathbf{q}_k} |\mathbf{G}_k|.$$

and

$$\begin{cases} \mathbf{G}_1^{-1} = \frac{\mathbf{a}_1^* \mathbf{a}_1^T}{\sigma^2} + (1 + |\alpha|^2) \mathbf{C}^{-1} \\ \mathbf{q}_1 = \frac{r_1 \mathbf{a}_1^*}{\sigma^2} + \alpha \mathbf{C}^{-1} \mathbf{h}_0. \end{cases}$$

$$\begin{cases} \mathbf{G}_k^{-1} = \frac{\mathbf{a}_k^* \mathbf{a}_k^T}{\sigma^2} + \mathbf{C}^{-1} + |\alpha|^2 (\mathbf{C}^{-1} - \mathbf{C}^{-1} \mathbf{G}_{k-1} \mathbf{C}^{-1}) \\ \mathbf{q}_k = \frac{r_k \mathbf{a}_k^*}{\sigma^2} + \alpha \mathbf{C}^{-1} \mathbf{G}_{k-1} \mathbf{q}_{k-1}. \end{cases}$$

$$\begin{cases} \mathbf{G}_N^{-1} = \frac{\mathbf{a}_N^* \mathbf{a}_N^T}{\sigma^2} + \mathbf{C}^{-1} - |\alpha|^2 \mathbf{C}^{-1} \mathbf{G}_{N-1} \mathbf{C}^{-1} \\ \mathbf{q}_N = \frac{r_N \mathbf{a}_N^*}{\sigma^2} + \alpha \mathbf{C}^{-1} \mathbf{G}_{N-1} \mathbf{q}_{N-1}. \end{cases}$$



## Design Methodology

- Let the code be with length  $N$  and rate  $R = K/N$ . The average block error rate can be upper-bounded by union bound as

$$\begin{aligned} P_e &\triangleq \Pr(\hat{\mathbf{x}} \neq \mathbf{x}) \\ &= 2^{-K} \sum_i \Pr(\hat{\mathbf{x}} \neq \mathbf{x}(i) | \mathbf{x}(i) \text{ is transmitted}) \\ &\leq 2^{-K} \sum_i \sum_{j \neq i} p_{j|i}, \end{aligned}$$

- The pairwise error probability

$$\begin{aligned}
p_{j|i} &= \Pr \left[ \log \frac{\Pr[\mathbf{r}|\mathbf{x}(j)]}{\Pr[\mathbf{r}|\mathbf{x}(i)]} > 0 \right] + \frac{1}{2} \Pr \left[ \log \frac{\Pr[\mathbf{r}|\mathbf{x}(j)]}{\Pr[\mathbf{r}|\mathbf{x}(i)]} = 0 \right]. \\
&= \Pr \left[ \log \frac{\prod_{k=1}^N e^{\mathbf{q}_k(j)^H \mathbf{G}_k(j) \mathbf{q}_k(j)} |\mathbf{G}_k(j)|}{\prod_{k=1}^N e^{\mathbf{q}_k(i)^H \mathbf{G}_k(i) \mathbf{q}_k(i)} |\mathbf{G}_k(i)|} > 0 \right] \\
&\quad + \frac{1}{2} \Pr \left[ \log \frac{\prod_{k=1}^N e^{\mathbf{q}_k(j)^H \mathbf{G}_k(j) \mathbf{q}_k(j)} |\mathbf{G}_k(j)|}{\prod_{k=1}^N e^{\mathbf{q}_k(i)^H \mathbf{G}_k(i) \mathbf{q}_k(i)} |\mathbf{G}_k(i)|} = 0 \right] \\
&= \Pr \left[ \sum_{k=1}^N \left( \mathbf{q}_k(j)^H \mathbf{G}_k(j) \mathbf{q}_k(j) - \mathbf{q}_k(i)^H \mathbf{G}_k(i) \mathbf{q}_k(i) \right) > \sum_{k=1}^N \left( \log \frac{|\mathbf{G}_k(i)|}{|\mathbf{G}_k(j)|} \right) \right] \\
&\quad + \frac{1}{2} \Pr \left[ \sum_{k=1}^N \left( \mathbf{q}_k(j)^H \mathbf{G}_k(j) \mathbf{q}_k(j) - \mathbf{q}_k(i)^H \mathbf{G}_k(i) \mathbf{q}_k(i) \right) = \sum_{k=1}^N \left( \log \frac{|\mathbf{G}_k(i)|}{|\mathbf{G}_k(j)|} \right) \right].
\end{aligned}$$

- Now we reform the  $\mathbf{q}$ -matrix and  $\mathbf{G}$ -matrix as

$$\mathbf{q}(j, i) = \begin{bmatrix} \mathbf{q}_1(j) \\ \mathbf{q}_2(j) \\ \vdots \\ \mathbf{q}_N(j) \\ \mathbf{q}_1(i) \\ \mathbf{q}_2(i) \\ \vdots \\ \mathbf{q}_N(i) \end{bmatrix}, \mathbf{G}(j, i) = \begin{bmatrix} \mathbf{G}_1(j) & 0 & \dots & 0 & 0 & \dots & 0 \\ 0 & \mathbf{G}_2(j) & \dots & 0 & 0 & \dots & 0 \\ \vdots & \vdots & \ddots & \vdots & \vdots & \dots & 0 \\ 0 & 0 & \dots & \mathbf{G}_N(j) & 0 & \dots & 0 \\ 0 & 0 & \dots & 0 & -\mathbf{G}_1(i) & \dots & 0 \\ \vdots & \vdots & \ddots & \vdots & \vdots & \ddots & 0 \\ 0 & 0 & \dots & 0 & 0 & \dots & -\mathbf{G}_N(i) \end{bmatrix}$$

$$p_{j|i} = \Pr \left[ \mathbf{q}^H(j, i) \mathbf{G}(j, i) \mathbf{q}(j, i) > \sum_{k=1}^N \left( \log \frac{|\mathbf{G}_k(i)|}{|\mathbf{G}_k(j)|} \right) \right] \\ + \frac{1}{2} \Pr \left[ \mathbf{q}^H(j, i) \mathbf{G}(j, i) \mathbf{q}(j, i) = \sum_{k=1}^N \left( \log \frac{|\mathbf{G}_k(i)|}{|\mathbf{G}_k(j)|} \right) \right],$$

where  $\mathbf{q}(j, i)$  is complex Gaussian with mean  $\mathbf{m}_q(j, i)$  and covariance  $\mathbf{S}_q(j, i)$ .

- Note that  $\mathbf{h}_k$  is complex Gaussian with mean  $\mu_k$  and covariance  $\mathbf{S}_k$ ,

$$\mu_k = E[\mathbf{h}_k] = E_{\mathbf{h}_{k-1}}[E_{\mathbf{h}_k}[\mathbf{h}_k | \mathbf{h}_{k-1}]] = E_{\mathbf{h}_{k-1}}[\alpha \mathbf{h}_{k-1}] = \alpha \mu_{k-1} = \alpha^k \mathbf{h}_0,$$

and

$$\begin{aligned} \mathbf{S}_k &= E[(\mathbf{h}_k - \mu_k)(\mathbf{h}_k - \mu_k)^H] \\ &= E[\mathbf{h}_k \mathbf{h}_k^H] - \mu_k E[\mathbf{h}_k^H] - E[\mathbf{h}_k] \mu_k^H + \mu_k \mu_k^H \\ &= E[(\alpha \mathbf{h}_{k-1} + \mathbf{v}_k)(\alpha \mathbf{h}_{k-1} + \mathbf{v}_k)^H] - |\alpha|^2 \mu_{k-1} E[\mathbf{h}_{k-1}^H] - |\alpha|^2 E[\mathbf{h}_{k-1}] \mu_{k-1}^H \\ &\quad + |\alpha|^2 \mu_{k-1} \mu_{k-1}^H \text{ (because } \mathbf{h}_k = \alpha \mathbf{h}_{k-1} + \mathbf{v}_k, \mu_k = \alpha \mu_{k-1}) \\ &= |\alpha|^2 E[\mathbf{h}_{k-1} \mathbf{h}_{k-1}^H] + E[\mathbf{v}_k \mathbf{v}_k^H] - |\alpha|^2 \mu_{k-1} E[\mathbf{h}_{k-1}^H] \\ &\quad - |\alpha|^2 E[\mathbf{h}_{k-1}] \mu_{k-1}^H + |\alpha|^2 \mu_{k-1} \mu_{k-1}^H \\ &= |\alpha|^2 E[(\mathbf{h}_{k-1} - \mu_{k-1})(\mathbf{h}_{k-1} - \mu_{k-1})^H] + \mathbf{C} \\ &= (1 + |\alpha|^2 + |\alpha|^4 + \dots + |\alpha|^{2(k-1)}) \mathbf{C} \\ &= \frac{1 - |\alpha|^{2k}}{1 - |\alpha|^2} \mathbf{C}. \end{aligned}$$

- Because  $|\alpha| < 1$ , the steady-state probability of  $\mathbf{h}_k$  (i.e.  $k \rightarrow \infty$ ) is

$$f(\mathbf{h}_k) \sim CN\left(0, \frac{1}{1 - |\alpha|^2} \mathbf{C}\right).$$

- Hence the mean vector of  $\mathbf{q}(j, i)$  is given as

$$\begin{aligned} \mathbf{m}_q(j, i) &= E[\mathbf{q}(j, i)] \\ &= \begin{bmatrix} \frac{E[r_1] \mathbf{a}_1^*(j)}{\sigma^2} + \alpha \mathbf{C}^{-1} E[\mathbf{h}_0] \\ \frac{E[r_2] \mathbf{a}_2^*(j)}{\sigma^2} + \alpha \mathbf{C}^{-1} \mathbf{G}_1(j) E[\mathbf{q}_1(j)] \\ \vdots \\ \frac{E[r_N] \mathbf{a}_N^*(j)}{\sigma^2} + \alpha \mathbf{C}^{-1} \mathbf{G}_{N-1}(j) E[\mathbf{q}_{N-1}(j)] \\ \frac{E[r_1] \mathbf{a}_1^*(i)}{\sigma^2} + \alpha \mathbf{C}^{-1} E[\mathbf{h}_0] \\ \frac{E[r_2] \mathbf{a}_2^*(i)}{\sigma^2} + \alpha \mathbf{C}^{-1} \mathbf{G}_1(i) E[\mathbf{q}_1(i)] \\ \vdots \\ \frac{E[r_N] \mathbf{a}_N^*(i)}{\sigma^2} + \alpha \mathbf{C}^{-1} \mathbf{G}_{N-1}(i) E[\mathbf{q}_{N-1}(i)] \end{bmatrix} \\ &= \mathbf{0}, \end{aligned}$$

- Covariance matrix of  $\mathbf{q}(j, i)$  is

$$\mathbf{S}_q(j, i) = E [\mathbf{q}(j, i)\mathbf{q}^H(j, i)],$$

where the  $(k, l)$  element in  $\mathbf{S}_q(j, i)$  is given as

$$\begin{aligned} & E[\mathbf{q}_k(n)\mathbf{q}_l^H(m)] \\ = & E \left[ \left( \frac{r_k \mathbf{a}_k^*(n)}{\sigma^2} + \alpha \mathbf{C}^{-1} \mathbf{G}_{k-1}(n) \mathbf{q}_{k-1}(n) \right) \left( \frac{r_l \mathbf{a}_l^*(m)}{\sigma^2} + \alpha \mathbf{C}^{-1} \mathbf{G}_{l-1}(m) \right. \right. \\ & \left. \left. \mathbf{q}_{l-1}(m) \right)^H \right] \\ = & \left( \mathbf{a}_k^T(n) E[\mathbf{h}_k \mathbf{h}_l^H] \mathbf{a}_l^*(m) + E[n_k n_l] \right) \frac{\mathbf{a}_k^*(n) \mathbf{a}_l^T(m)}{\sigma^4} \\ & + \frac{\alpha}{\sigma^2} \mathbf{C}^{-1} \mathbf{G}_{k-1}(n) E[\mathbf{q}_{k-1}(n) \mathbf{h}_l^H] \mathbf{a}_l^*(m) \mathbf{a}_l^T(m) \\ & + \frac{\alpha^*}{\sigma^2} \mathbf{a}_k^*(n) \mathbf{a}_k^T(n) E[\mathbf{h}_k \mathbf{q}_{l-1}^H(m)] \mathbf{G}_{l-1}(m) \mathbf{C}^{-1} \\ & + |\alpha|^2 \mathbf{C}^{-1} \mathbf{G}_{k-1}(n) E[\mathbf{q}_{k-1}(n) \mathbf{q}_{l-1}^H(m)] \mathbf{G}_{l-1}(m) \mathbf{C}^{-1}. \end{aligned}$$

- Since  $\mathbf{S}_q(j, i)$  is real and symmetric, it can be represented by two real and symmetric square roots

$$\mathbf{S}_q(j, i) = \mathbf{S}_q^{1/2}(j, i)\mathbf{S}_q^{1/2}(j, i).$$

- And  $\mathbf{S}_q^{1/2}(j, i)\mathbf{G}(j, i)\mathbf{S}_q^{1/2}(j, i)$  is also real and symmetric, it can be expressed by its eigenvalues  $\lambda_n$  and eigenvectors  $\mathbf{k}_n$  as

$$\mathbf{S}_q^{1/2}(j, i)\mathbf{G}(j, i)\mathbf{S}_q^{1/2}(j, i) = \sum_{n=1}^{2MN} \lambda_n \mathbf{k}_n \mathbf{k}_n^T,$$

- So,  $\mathbf{q}^H(j, i)\mathbf{G}(j, i)\mathbf{q}(j, i)$  can be specified as

$$\begin{aligned}
 & \left(\mathbf{S}_q^{-1/2}(j, i)\mathbf{q}(j, i)\right)^H \mathbf{S}_q^{1/2}(j, i)\mathbf{G}(j, i)\mathbf{S}_q^{1/2}(j, i) \left(\mathbf{S}_q^{-1/2}(j, i)\mathbf{q}(j, i)\right) \\
 &= \sum_{n=1}^{2MN} \lambda_n |\mathbf{k}_n^T \mathbf{S}_q^{-1/2}(j, i)\mathbf{q}(j, i)|^2 \\
 &= \sum_{n=1}^{2MN} \lambda_n |X_n|^2,
 \end{aligned}$$

where  $X_n = \mathbf{k}_n^T \mathbf{S}_q^{-1/2}(j, i)\mathbf{q}(j, i)$  is complex Gaussian with mean

$$\mathbf{k}_n^T \mathbf{S}_q^{-1/2}(j, i) E[\mathbf{q}(j, i)] = \mathbf{0},$$

variance

$$\mathbf{k}_n^T \mathbf{S}_q^{-1/2}(j, i) E[\mathbf{q}(j, i)\mathbf{q}(j, i)^H] \mathbf{S}_q^{-1/2}(j, i)\mathbf{k}_n = \mathbf{k}_n^T \mathbf{k}_n = 1,$$

and is independent in  $n$ .



$$\begin{aligned}
p_{j|i} &= \Pr \left[ \mathbf{q}^H(j, i) \mathbf{G}(j, i) \mathbf{q}(j, i) > \sum_{k=1}^n \left( \log \frac{|\mathbf{G}_k(i)|}{|\mathbf{G}_k(j)|} \right) \right] \\
&= \Pr \left[ \sum_{l=1}^{\bar{L}} \bar{\lambda}_l \chi^2(2\kappa_l) > \sum_{k=1}^n \left( \log \frac{|\mathbf{G}_k(i)|}{|\mathbf{G}_k(j)|} \right) \right]
\end{aligned}$$

- Assuming the eigenvalues have been ordered as  $\bar{\lambda}_1 > \bar{\lambda}_2 > \dots > \bar{\lambda}_p > 0 > \bar{\lambda}_{p+1} > \bar{\lambda}_{p+2} > \dots > \bar{\lambda}_{\bar{L}}$ , and  $n = \sum_{l=1}^{\bar{L}} \kappa_l$ , a closed-form expression of  $p_{j|i}$  is

$$\begin{aligned}
p_{j|i} &= \Pr \left[ \mathbf{q}^H(j, i) \mathbf{G}(j, i) \mathbf{q}(j, i) > \sum_{k=1}^n \left( \log \frac{|\mathbf{G}_k(i)|}{|\mathbf{G}_k(j)|} \right) \right] \\
&= \sum_{l=1}^p \frac{1}{(\kappa_l - 1)!} \left[ \frac{\partial^{\kappa_l - 1}}{\partial x^{\kappa_l - 1}} F_l(x) \right]_{x=\bar{\lambda}_l}
\end{aligned}$$

where

$$F_l(x) = x^{q-1} \exp \left[ \frac{-\sum_{k=1}^n \left( \log \frac{|\mathbf{G}_k(i)|}{|\mathbf{G}_k(j)|} \right)}{2x} \right] \prod_{r=1, r \neq l}^{\bar{L}} (x - \bar{\lambda}_r)^{-\kappa_r}.$$

- Define the energy function of simulated annealing algorithm as

$$E = 2^{-K} \sum_{i=1}^{2^K} \sum_{j=1, j \neq i}^{2^K} p_{j|i},$$

which results in searching for a code having minimum average block error rate.

- The perturbation of  $u_{th}$  codeword will only affect the pairwise error probabilities of those  $u_{th}$ -related pairs, i.e.  
 $(1, u), (2, u), \dots, (u-1, u), (u+1, u), \dots, (2^k, u)$  and  
 $(u, 1), (u, 2), \dots, (u, u-1), (u, u+1), \dots, (u, 2^k)$ .

- The energy function is now defined as

$$E(C') = E(C) - 2^{-K} \sum_{i=1 \dots 2^K, i \neq u, (i,u) \in C} (p_{i|u} + p_{u|i})$$

$$+ 2^{-K} \sum_{i'=1 \dots 2^K, i' \neq u', (i',u') \in C'} (p_{i'|u'} + p_{u'|i'}).$$

- It saves about  $1 - \frac{2 \cdot (2^K - 1)}{2^K \cdot (2^K - 1)} = 1 - 2^{-K+1}$  portion of running time, but the cost is that all pairwise error probabilities must be recorded.

## Simulation Results

- The codes we found by our criterion and by [7]'s were first encoded as BPSK signals, suffered fading through the Gauss-Markov channel, and then decoded according to the optimal rule

$$\hat{\mathbf{a}}(\mathbf{r}) = \mathbf{a}(i'), \text{ where } i' = \arg \max_{i \in 2^K} \left( \sum_{k=1}^N (\mathbf{q}_k(i)^H \mathbf{G}_k(i) \mathbf{q}_k(i) + \log |\mathbf{G}_k(i)|) \right).$$

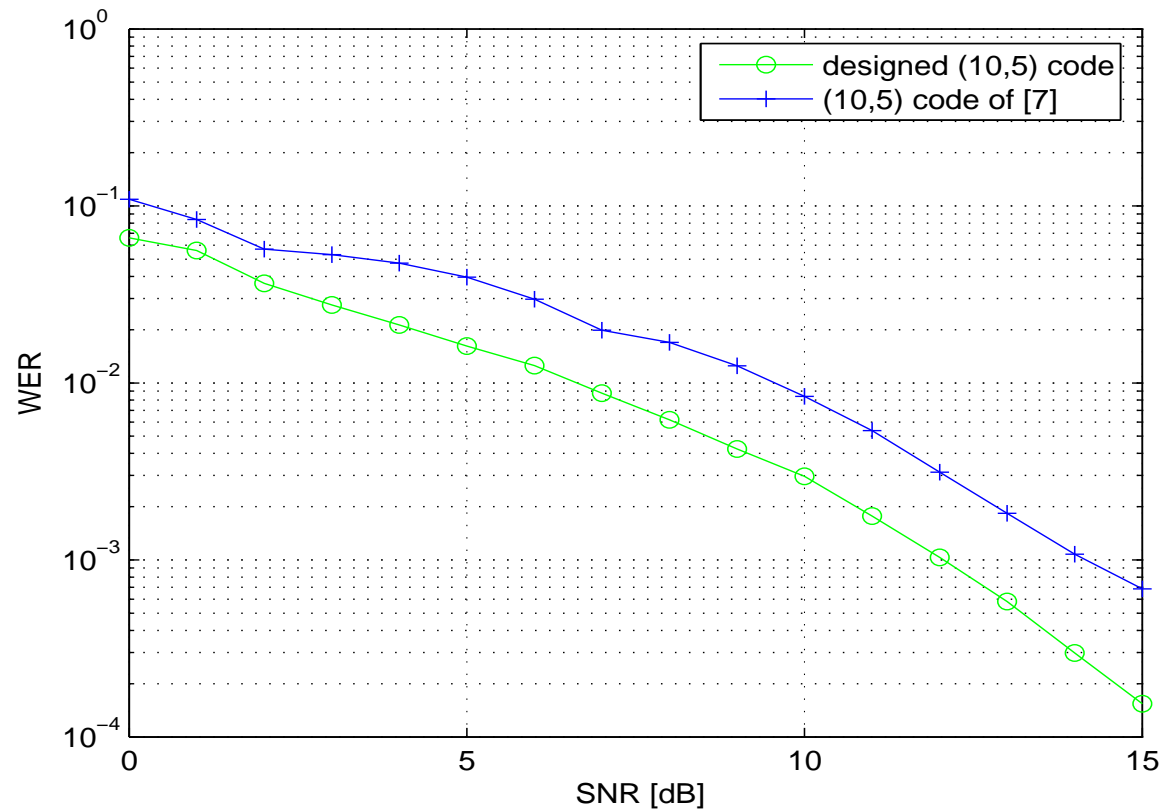


Figure 1: Word error rate for (10,5) code on the Gauss-Markov channel ( $M=1$ ) with initial channel coefficient  $h_0$  assumed to be known at transmitter and receiver is given as 1, and  $h_k = 0.9h_{k-1} + v_k$ , where  $v_k$  has mean  $d=0$  and covariance  $C = 0.001$ .

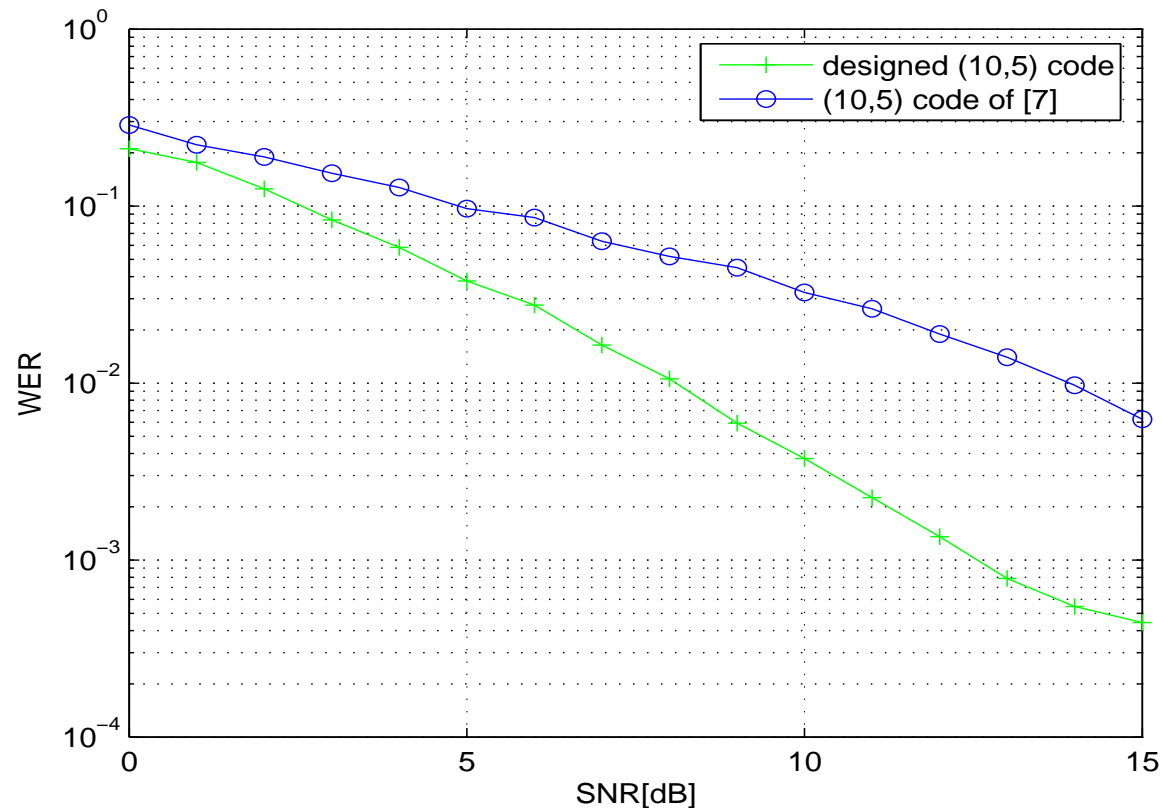


Figure 2: Word error rate for (10,5) code on the Gauss-Markov channel ( $M=2$ ) with  $\mathbf{h}_0$  assumed to be known at transmitter and receiver is given as  $\mathbf{1}$ , and  $\mathbf{h}_k = 0.9\mathbf{h}_{k-1} + \mathbf{v}_k$ , where  $\mathbf{v}_k$  has mean  $\mathbf{d}=\mathbf{0}$  and covariance  $\mathbf{C}=0.001\mathbf{I}$ .

Table 1: comparison of complexity

| Criterion for Gauss-Markov channel          |                                                           |                    | Criterion of [7]                         |                          |                    |
|---------------------------------------------|-----------------------------------------------------------|--------------------|------------------------------------------|--------------------------|--------------------|
|                                             | number of multiplication                                  | number of addition |                                          | number of multiplication | number of addition |
| $\mathbf{Q}(j, i)$                          | $24N-16$                                                  | $8N-6$             | $\mathbf{P}_B(i)$                        | $2N^2+12N+13$            | $N^2+8N+4$         |
| $\mathbf{G}(j, i)$                          | $54N-40$                                                  | $18N-12$           | $\mathbf{Q}(j, i)$                       | $0$                      | $N^2+2N+1$         |
| $\mathbf{S}_q(j, i)$                        | $\frac{4}{3}N^3 + \frac{187}{6}N^2 - \frac{241}{6}N + 23$ | $4N^2+8N+2$        | $\mathbf{S}_y(i)$                        | $2N^2+9N+7$              | $2N^2+6N+4$        |
| $\mathbf{S}_q(j, i) \cdot \mathbf{G}(j, i)$ | $64N^3$                                                   | $64N^3 - 16N^2$    | $\mathbf{S}_y(i) \cdot \mathbf{Q}(j, i)$ | $(N+1)^3$                | $N(N+1)^2$         |

- For  $N = 10$ , the total numbers of multiplications and additions are 68795 and 63124 respectively for our criterion, and 1961 and 1779 respectively for that given in [7].

## Conclusion

- We describe the Gauss-Markov channel and its likelihood function which is a recursive form of every prior state.
- Then we derive the pairwise error probability as a function of the criterion for the simulated annealing algorithm.
- Our designed codes provide a coding gain of about 3.5 dB and 6 dB on the Gauss-Markov channels with channel lengths 1 and 2 respectively over those given in [7] at  $WER = 10^{-2}$ .